### Low-Dropout, 120mA Linear Regulators

#### **General Description**

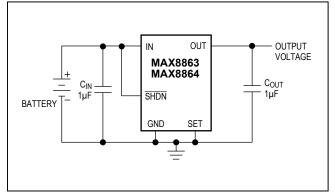
The MAX8863T/S/R and MAX8864T/S/R low-dropout linear regulators operate from a +2.5V to +6.5V input range and deliver up to 120mA. A PMOS pass transistor allows the low,  $80\mu A$  supply current to remain independent of load, making these devices ideal for battery-operated portable equipment such as cellular phones, cordless phones, and modems.

The devices feature Dual Mode™ operation: their output voltage is preset (at 3.15V for the T versions, 2.84V for the S versions, or 2.80V for the R versions) or can be adjusted with an external resistor divider. Other features include low-power shutdown, short-circuit protection, thermal shutdown protection, and reverse battery protection. The MAX8864 also includes an auto-discharge function, which actively discharges the output voltage to ground when the device is placed in shutdown mode. Both devices come in a miniature 5-pin SOT23 package.

#### **Applications**

- Cordless Telephones
- PCS Telephones
- Cellular Telephones
- PCMCIA Cards
- Modems
- Hand-Held Instruments
- Palmtop Computers
- Electronic Planners

### **Typical Operating Circuit**



Dual Mode is a trademark of Maxim Integrated Products, Inc.

#### **Benefits and Features**

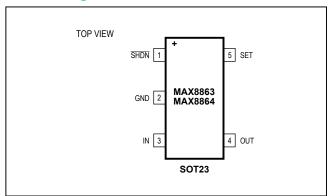
- Low Cost
- Low, 55mV Dropout Voltage at 50mA I<sub>OUT</sub>
- Low, 68µA No-Load Supply Current Low, 80µA Operating Supply Current (even in dropout)
- Low, 350μV<sub>RMS</sub> Output Noise
- Miniature External Components
- Thermal Overload Protection
- Output Current Limit
- Reverse Battery Protection
- Dual Mode Operation: Fixed or Adjustable (1.25V to 6.5V) Output
- Low-Power Shutdown

#### **Ordering Information**

PART	TEMP RANGE	PIN-PACKAGE
MAX8863TEUK+T	-40°C to +85°C	5 SOT23
MAX8863TMUK/PR3+T	-55°C to +125°C	5 SOT23
MAX8863SEUK+T	-40°C to +85°C	5 SOT23
MAX8863REUK+T	-40°C to +85°C	5 SOT23
MAX8864TEUK+T	-40°C to +85°C	5 SOT23
MAX8864SEUK+T	-40°C to +85°C	5 SOT23
MAX8864REUK+T	-40°C to +85°C	5 SOT23

\*Alternate marking information: CY\_\_ = MAX8863T, CZ\_\_ = MAX8863S, DA\_\_ = MAX8864T, DB\_\_ = MAX8864S +Denotes a lead(Pb)-free/RoHS-compliant package.

### **Pin Configuration**





### Low-Dropout, 120mA Linear Regulators

### **Absolute Maximum Ratings**

V <sub>IN</sub> to GND7V to +7V	C
Output Short-Circuit Duration	C
SET to GND0.3V to +7V	
SHDN to GND7V to +7V	J
SHDN to IN7V to +0.3V	S
OUT to GND0.3V to (V <sub>IN</sub> + 0.3V)	L
Continuous Power Dissipation (multilayer board, $T_A = +70$ °C)	S
SOT23 (3.9mW/°C above +70°C)312.6mW	

Operating Temperature Range	40°C to +85°C
Operating Temperature	
Range (MAX8863TMUK/PR3+)	55°C to +125°C
Junction Temperature	+150°C
Storage Temperature Range	65°C to +150°C
Lead Temperature (soldering, 10s)	+300°C
Soldering Temperature (reflow)	+260°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

#### **Package Thermal Characteristics (Note 1)**

SOT23

Junction-to-Ambient Thermal Resistance ( $\theta_{JA}$ ) .....255.9°C/W Junction-to-Case Thermal Resistance ( $\theta_{JC}$ )......81°C/W

Note 1: Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a four-layer board. For detailed information on package thermal considerations, refer to www.maximintegrated.com/thermal-tutorial.

#### **Electrical Characteristics**

 $(V_{IN} = +3.6V, V_{GND} = 0V, T_A = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted. Typical values are at } T_A = +25^{\circ}\text{C.})$  (Note 2)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
Input Voltage (Note 3)	V <sub>IN</sub>			2.5		6.5	V
			MAX886_T	3.05	3.15	3.25	
Output Voltage	V <sub>OUT</sub>	0mA ≤ I <sub>OUT</sub> ≤ 50mA, SET = GND	MAX886_S	2.75	2.84	2.93	_ v
		OOMINA, GET OND	MAX886_R	2.70	2.80	2.88	
Adjustable Output Voltage Range (Note 4)	V <sub>OUT</sub>			V <sub>SET</sub>		6.5	V
Maximum Output Current				120			mA
Current Limit (Note 5)	I <sub>LIM</sub>				280		mA
0 15: 0 1	IQ	SET = GND	I <sub>LOAD</sub> = 0mA		68	150	μА
Ground Pin Current			I <sub>LOAD</sub> = 50mA		80		
Drangut Valtage (Note 6)		I <sub>OUT</sub> = 1mA			1.1		\/
Dropout Voltage (Note 6)		I <sub>OUT</sub> = 50mA			55	120	mV
Line Regulation	ΔV <sub>LNR</sub>	V <sub>IN</sub> = 2.5V to 6.5V, SET tied to OUT, I <sub>OUT</sub> = 1mA		-0.15	0	+0.15	%/V
Load Degulation	437	I <sub>OUT</sub> = 0mA to 50mA	SET = GND		0.011	0.040	%/mA
Load Regulation	$\Delta V_{LDR}$		SET tied to OUT		0.006		
Output Voltage Noise		10Hz to 1MHz	C <sub>OUT</sub> = 1µF		350		
			C <sub>OUT</sub> = 100μF		220		μV <sub>RMS</sub>
SHUTDOWN							
SUDN Input Throshold	V <sub>IH</sub>			2.0			V
SHDN Input Threshold	V <sub>IL</sub>					0.4	] <b>'</b>

#### **Electrical Characteristics (continued)**

 $(V_{IN} = +3.6V, V_{GND} = 0V, T_A = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted. Typical values are at } T_A = +25^{\circ}\text{C.})$  (Note 2)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
SUDN Input Dice Current		\/	T <sub>A</sub> = +25°C		0	100	n 1
SHDN Input Bias Current	ISHDN	V <sub>SHDN</sub> = V <sub>IN</sub>	$T_A = T_{MAX}$		0.05		nA
Shutdown Supply Current		V <sub>OUT</sub> = 0V	T <sub>A</sub> = +25°C		0.0001	1	μА
Shutdown Supply Current	IQSHDN		$T_A = T_{MAX}$		0.02		
Shutdown Discharge Resistance (MAX8864)					300		Ω
SET INPUT							
			T <sub>A</sub> = +25°C	1.225	1.25	1.275	
Set Reference Voltage (Note 4)	$V_{SET}$ $V_{IN} = 2.5V to I_{OUT} = 1 mA$	$V_{IN} = 2.5V \text{ to } 6.5V,$	E temp range	1.215	1.25	1.285	V
(1010-1)		1001	M temp range	1.205	1.25	1.285	
Set Input Leakage Current	1	\/ = 1.2\/	T <sub>A</sub> = +25°C		0.015	2.5	nA
(Note 4)	I <sub>SET</sub> V <sub>SET</sub>	V <sub>SET</sub> = 1.3V	$T_A = T_{MAX}$		0.5		
THERMAL PROTECTION	·						
Thermal Shutdown Temperature	T <sub>SHDN</sub>				170		°C
Thermal Shutdown Hysteresis	ΔT <sub>SHDN</sub>				20		°C

Note 2: Limits are 100% production tested at  $T_A = +25$ °C. Limits over the operating temperature range are guaranteed through correlation using Statistical Quality Control (SQC) Methods.

Note 3: Guaranteed by line regulation test.

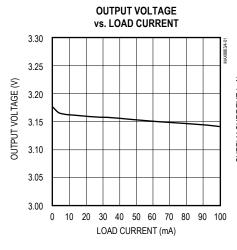
Note 4: Adjustable mode only.

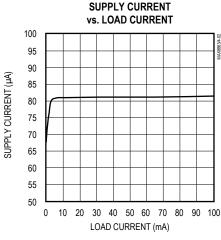
Note 5: Not tested. For design purposes, the current limit should be considered 120mA minimum to 420mA maximum.

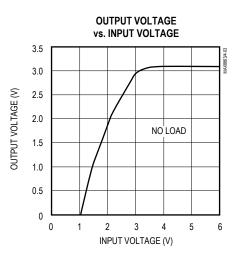
Note 6: The dropout voltage is defined as (V<sub>IN</sub> - V<sub>OUT</sub>) when V<sub>OUT</sub> is 100mV below the value of V<sub>OUT</sub> for V<sub>IN</sub> = V<sub>OUT</sub> +2V.

### **Typical Operating Characteristics**

 $(V_{IN} = +3.6V, C_{IN} = 1\mu F, C_{OUT} = 1\mu F, T_A = +25^{\circ}C, MAX886_T, unless otherwise noted.)$ 

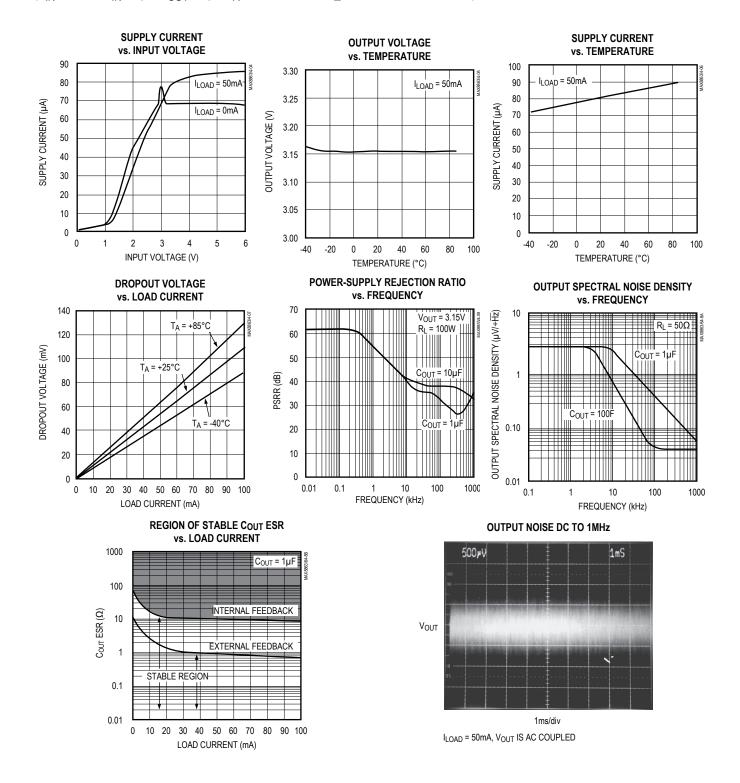






#### **Typical Operating Characteristics (continued)**

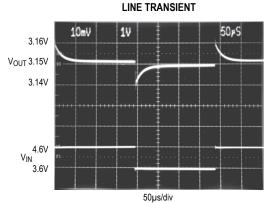
 $(V_{IN}$  = +3.6V,  $C_{IN}$  = 1 $\mu$ F,  $C_{OUT}$  = 1 $\mu$ F,  $T_A$  = +25°C, MAX886\_T, unless otherwise noted.)



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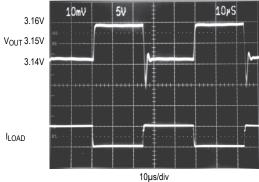
### **Typical Operating Characteristics (continued)**

(V<sub>IN</sub> = +3.6V, C<sub>IN</sub> = 1 $\mu$ F, C<sub>OUT</sub> = 1 $\mu$ F, T<sub>A</sub> = +25°C, MAX886\_T, unless otherwise noted.)



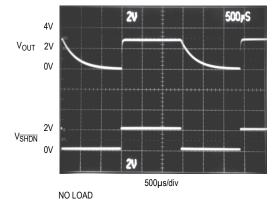
 $I_{LOAD}$  = 50mA,  $V_{OUT}$  IS AC COUPLED

### LOAD TRANSIENT

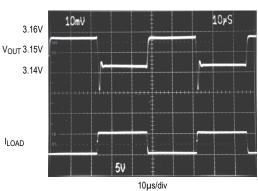


 $V_{IN}$  =  $V_{OUT}$  + 0.2V,  $I_{LOAD}$  = 0mA to 50mA,  $C_{IN}$  =  $10 \, \mu F,$   $V_{OUT}$  IS AC COUPLED

#### MAX8864 SHUTDOWN (NO LOAD)

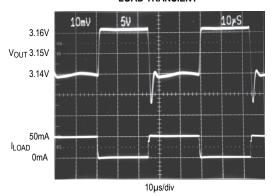


#### LOAD TRANSIENT



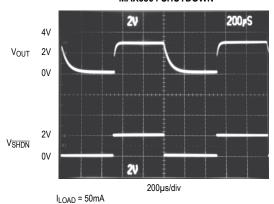
 $I_{LOAD}$  = 0mA to 50mA,  $C_{IN}$  = 10 $\mu$ F,  $V_{OUT}$  IS AC COUPLED

#### LOAD TRANSIENT



 $V_{IN}$  =  $V_{OUT}$  + 0.1V,  $I_{LOAD}$  = 0mA to 50mA,  $C_{IN}$  =  $10\mu F,$   $V_{OUT}$  IS AC COUPLED

#### MAX8864 SHUTDOWN



#### **Pin Description**

PIN	NAME	FUNCTION
1	SHDN	Active-Low Shutdown Input. A logic low reduces the supply current to 0.1nA. On the MAX8864, a logic low also causes the output voltage to discharge to GND. Connect to IN for normal operation.
2	GND	Ground. This pin also functions as a heatsink. Solder to large pads or the circuit board ground plane to maximize thermal dissipation.
3	IN	Regulator Input. Supply voltage can range from +2.5V to +6.5V. Bypass with 1µF to GND (see <i>Capacitor Selection</i> and <i>Regulator Stability</i> ).
4	OUT	Regulator Output. Fixed or adjustable from +1.25V to +6.5V. Sources up to 120mA. Bypass with a $1\mu F$ , < $0.2\Omega$ typical ESR capacitor to GND.
5	SET	Feedback Input for Setting the Output Voltage. Connect to GND to set the output voltage to the preset 2.80V (MAX886_R), 2.84V (MAX886_S), or 3.15V (MAX886_T). Connect to an external resistor-divider for adjustable-output operation.

#### **Detailed Description**

The MAX8863/MAX8864 are low-dropout, low-quiescent-current linear regulators designed primarily for battery-powered applications. They supply an adjustable 1.25V to 6.5V output or a preselected 2.80V (MAX886\_R), 2.84V (MAX886\_S), or 3.15V (MAX886\_T) output for load currents up to 120mA. As illustrated in Figure 1, these devices consist of a 1.25V reference, error amplifier, MOSFET driver, P-channel pass transistor, Dual Mode  $^{\rm TM}$  comparator, and internal feedback voltage divider.

The 1.25V bandgap reference is connected to the error amplifier's inverting input. The error amplifier compares this reference with the selected feedback voltage and amplifies the difference. The MOSFET driver reads the error signal and applies the appropriate drive to the

P-channel pass transistor. If the feedback voltage is lower than the reference, the pass-transistor gate is pulled lower, allowing more current to pass and increasing the output voltage. If the feedback voltage is too high, the pass-transistor gate is pulled up, allowing less current to pass to the output.

The output voltage is fed back through either an internal resistor voltage divider connected to the OUT pin, or an external resistor network connected to the SET pin. The Dual Mode comparator examines the SET voltage and selects the feedback path. If SET is below 60mV, internal feedback is used and the output voltage is regulated to the preset output voltage. Additional blocks include a current limiter, reverse battery protection, thermal sensor, and shutdown logic.

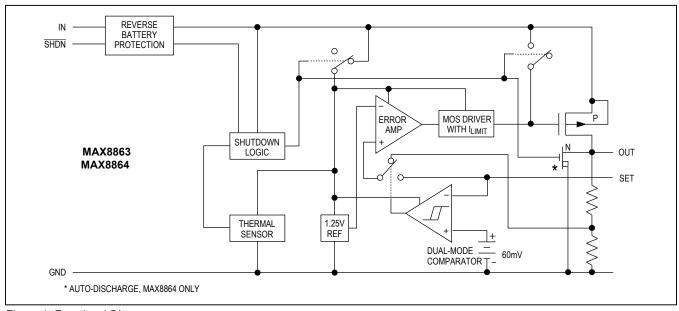


Figure 1. Functional Diagram

### Low-Dropout, 120mA Linear Regulators

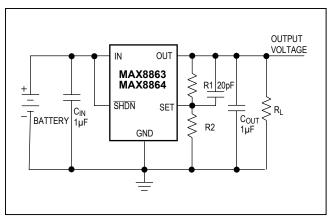


Figure 2. Adjustable Output Using External Feedback Resistors

#### **Internal p-Channel Pass Transistor**

The MAX8863/MAX8864 feature a  $1.1\Omega$  typical pMOSFET pass transistor. This provides several advantages over similar designs using PNP pass transistors, including longer battery life.

The pMOSFET requires no base drive current, which reduces quiescent current considerably. PNP-based regulators waste considerable amounts of current in dropout when the pass transistor saturates. They also use high base-drive currents under large loads. The MAX8863/MAX8864 do not suffer from these problems, and consume only  $80\mu A$  of quiescent current, whether in dropout, light load, or heavy load applications (see *Typical Operating Characteristics*).

#### **Output Voltage Selection**

The MAX8863/MAX8864 feature Dual Mode operation: they operate in either a preset voltage mode or an adjustable mode.

In preset voltage mode, internal, trimmed feedback resistors set the MAX886\_R output to 2.80V, the MAX886\_S output to 2.84V, and the MAX886\_T output to 3.15V. Select this mode by connecting SET to ground.

In adjustable mode, select an output between 1.25V and 6.5V using two external resistors connected as a voltage divider to SET (Figure 2). The output voltage is set by the following equation:

$$V_{OUT} = V_{SET} (1 + R1 / R2)$$

where  $V_{SFT}$  = 1.25V. To simplify resistor selection:

$$R1 = R2 \left( \frac{V_{OUT}}{V_{SFT}} - 1 \right)$$

Choose R2 =  $100k\Omega$  to optimize power consumption, accuracy, and high-frequency power-supply rejection. The total current through the external resistive feedback and load resistors should not be less than  $10\mu A$ . Since the V<sub>SET</sub> tolerance is typically less than  $\pm 25 mV$ , the output can be set using fixed resistors instead of trim pots. Connect a 10pF to 25pF capacitor across R1 to compensate for layout-induced parasitic capacitances.

In preset voltage mode, impedances between SET and ground should be less than  $100k\Omega$ . Otherwise, spurious conditions could cause the voltage at SET to exceed the 60mV Dual Mode threshold.

#### **Shutdown**

A low input on the SHDN pin shuts down the MAX8863/MAX8864. In shutdown mode, the pass transistor, control circuit, reference, and all biases are turned off, reducing the supply current to typically 0.1nA. Connect SHDN to IN for normal operation. The MAX8864 output voltage is actively discharged to ground when the part is placed in shutdown (see *Typical Operating Characteristics*).

#### **Current Limit**

The MAX8863/MAX8864 include a current limiter that monitors and controls the pass transistor's gate voltage, estimating the output current and limiting it to about 280mA. For design purposes, the current limit should be considered 120mA (min) to 420mA (max). The output can be shorted to ground for an indefinite time period without damaging the part.

#### **Thermal Overload Protection**

Thermal overload protection limits total power dissipation in the MAX8863/MAX8864. When the junction temperature exceeds  $T_J$  = +170°C, the thermal sensor sends a signal to the shutdown logic, turning off the pass transistor and allowing the IC to cool. The thermal sensor will turn the pass transistor on again after the IC's junction temperature typically cools by 20°C, resulting in a pulsed output during continuous thermal overload conditions.

Thermal overload protection is designed to protect the MAX8863/MAX8864 in the event of fault conditions. Stressing the device with high load currents and high input-output differential voltages (which result in die temperatures above +125°C) may cause a momentary overshoot (2% to 8% for 200ms) when the load is completely removed. This can be remedied by raising the minimum load current from 0 $\mu$ A (+125°C) to 100 $\mu$ A (+150°C). For continuous operation, do not exceed the absolute maximum junction temperature rating of T<sub>J</sub> = +150°C.

### Low-Dropout, 120mA Linear Regulators

#### **Operating Region and Power Dissipation**

Maximum power dissipation of the MAX8863/MAX8864 depends on the thermal resistance of the case and circuit board, the temperature difference between the die junction and ambient air, and the rate of air flow. The power dissipation across the device is  $P = I_{OUT} (V_{IN} - V_{OUT})$ . The resulting maximum power dissipation is:

$$P_{MAX} = (T_J - T_A) / \theta_{JA}$$

where (T<sub>J</sub> - T<sub>A</sub>) is the temperature difference between the MAX8863/MAX8864 die junction and the surrounding air, and  $\theta_{JA}$  is the thermal resistance of the chosen package to the surrounding air.

The GND pin of the MAX8863/MAX8864 performs the dual function of providing an electrical connection to ground and channeling heat away. Connect the GND pin to ground using a large pad or ground plane.

#### **Reverse Battery Protection**

The MAX8863/MAX8864 have a unique protection scheme that limits the reverse supply current to less than 1mA when either  $V_{\text{IN}}$  or  $V_{\overline{\text{SHDN}}}$  falls below ground. The circuitry monitors the polarity of these two pins, disconnecting the internal circuitry and parasitic diodes when the battery is reversed. This feature prevents the device from overheating and damaging the battery.

#### V<sub>IN</sub> > 5.5V Minimum Load Current

When operating the MAX8863/MAX8864 with an input voltage above 5.5V, a minimum load current of  $20\mu A$  is required to maintain regulation in preset voltage mode. When setting the output with external resistors, the minimum current through the external feedback resistors and load must be  $30\mu A$ .

#### **Applications Information**

#### **Capacitor Selection and Regulator Stability**

Normally, use a  $1\mu F$  capacitor on the input and a  $1\mu F$  capacitor on the output of the MAX8863/MAX8864. Larger input capacitor values and lower ESR provide better supply-noise rejection and transient response. A higher-value input capacitor ( $10\mu F$ ) may be necessary if large, fast transients are anticipated and the device is

located several inches from the power source. Improve load-transient response, stability, and power-supply rejection by using large output capacitors. For stable operation over the full temperature range, with load currents up to 120mA, a minimum of 1µF is recommended.

#### Noise

The MAX8863/MAX8864 exhibit  $350\mu V_{RMS}$  noise during normal operation. When using the MAX8863/MAX8864 in applications that include analog-to-digital converters of greater than 12 bits, consider the ADC's power-supply rejection specifications (see the Output Noise DC to 1MHz photo in the *Typical Operating Characteristics*).

## Power-Supply Rejection and Operation from Sources Other than Batteries

The MAX8863/MAX8864 are designed to deliver low dropout voltages and low quiescent currents in battery-powered systems. Power-supply rejection is 62dB at low frequencies and rolls off above 300Hz. As the frequency increases above 20kHz, the output capacitor is the major contributor to the rejection of power-supply noise (see the Power-Supply Rejection Ratio vs. Ripple Frequency graph in the *Typical Operating Characteristics*).

When operating from sources other than batteries, improve supply-noise rejection and transient response by increasing the values of the input and output capacitors, and using passive filtering techniques (see the supply and load-transient responses in the <u>Typical Operating Characteristics</u>).

#### **Load Transient Considerations**

The MAX8863/MAX8864 load-transient response graphs (see *Typical Operating Characteristics*) show two components of the output response: a DC shift of the output voltage due to the different load currents, and the transient response. Typical overshoot for step changes in the load current from 0mA to 50mA is 12mV. Increasing the output capacitor's value and decreasing its ESR attenuates transient spikes.

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### Low-Dropout, 120mA Linear Regulators

#### Input-Output (Dropout) Voltage

A regulator's minimum input-output voltage differential (or dropout voltage) determines the lowest usable supply voltage. In battery-powered systems, this will determine the useful end-of-life battery voltage. Because the MAX8863/ MAX8864 use a pMOSFET pass transistor, their dropout voltage is a function of  $R_{DS(ON)}$  multiplied by the load current (see *Electrical Characteristics*).

### **Package Information**

For the latest package outline information and land patterns (footprints), go to <a href="www.maximintegrated.com/packages">www.maximintegrated.com/packages</a>. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

PACKAGE PACKAGE CODE		OUTLINE NO.	LAND PATTERN NO.
5 SOT23	U5+1	21-0057	90-0174

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### Low-Dropout, 120mA Linear Regulators

### **Revision History**

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
3	5/11	Added lead-free designation and updated continuous power dissipation and $\theta_{\mbox{\scriptsize JA}}$ specs	1, 2
4	7/12	Corrected units for TOC 7	4
5	4/13	Updated Ordering Information and Electrical Characteristics tables	1, 3
6	5/16	Updating Ordering Information table	1

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